## Golub-Kahan iterative bidiagonalization and stopping criteria in ill-posed problems

Iveta Hnětynková *,**, Martin Plešinger *, Zdeněk Strakoš *,**<br>* Academy of Sciences of the Czech republic, Institute of Computer Science, Prague<br>** Charles University, Faculty of Mathematics and Physics, Department of Numerical Mathematics, Prague

## Outline

1. Problem formulation
2. Regularization by Golub-Kahan bidiagonalization
3. How to identify the noise
4. Summary and future work

## 1. Problem formulation

Consider an ill-posed linear system

$$
A x \approx b, \quad A \in \mathbb{R}^{n \times m}, \quad b \in \mathbb{R}^{n},
$$

with a noise contaminated right-hand side

$$
b=b^{\text {exact }}+b^{\text {noise }} \neq 0 \in \mathbb{R}^{n}, \quad\left\|b^{\text {exact }}\right\| \gg\left\|b^{\text {noise }}\right\|
$$

Possible difficulties:

- the noise component $b^{\text {noise }}$ is unknown;
- the rank of $A$ is not well defined (singular values of $A$ decay gradually to zero);
- the solution is sensitive on small perturbations in data.

Denote $l=\operatorname{rank}(A)$. Consider the singular value decomposition

$$
\begin{gathered}
A=\tilde{U} \tilde{\Sigma} \tilde{V}^{T}=\sum_{i=1}^{l} \tilde{u}_{i} \tilde{\sigma}_{i} \tilde{v}_{i}^{T} \\
\tilde{U}=\left[\tilde{u}_{1}, \ldots, \tilde{u}_{l}\right], \quad \tilde{V}=\left[\tilde{v}_{1}, \ldots, \tilde{v}_{l}\right], \quad \tilde{\Sigma}=\operatorname{diag}\left(\tilde{\sigma}_{1}, \ldots, \tilde{\sigma}_{l}\right)
\end{gathered}
$$

The least squares method (LS) minimizes $\|b-A x\|$ and

$$
\begin{aligned}
x^{L S} & =\sum_{i=1}^{l} \frac{\tilde{u}_{i}^{T} b}{\tilde{\sigma}_{i}} \tilde{v}_{i} \\
& =\sum_{i=1}^{l} \frac{\tilde{u}_{i}^{T} b^{e x a c t}}{\tilde{\sigma}_{i}} \tilde{v}_{i}+\sum_{i=1}^{l} \frac{\tilde{u}_{i}^{T} b^{n o i s e}}{\tilde{\sigma}_{i}} \tilde{v}_{i}
\end{aligned}
$$

Thus components of the solution corresponding to small singular values may be dominated by errors in $b$, the solution is meaningless.

Regularization methods are used to suppress the effect of errors in the data and extract the essential information about the system, e.g.,

- truncated SVD, truncated total least squares, Tikhonov reguIarization, see [Hansen, O'Leary - 97], [Fierro, Golub, Hansen, O'Leary - 97], [Hansen - 98], [Golub, Hansen, O'Leary - 99], [Sima, Van Huffel, Golub - 04], [Kilmer, Hansen, Espanol - 06],
- methods based on iterative Golub-Kahan bidiagonalization as LSQR, hybrid methods, see [Paige, Saunders - 82], [Bjorck - 96], [Hansen - 97], [Hanke - 01], ...


## 2. Regularization by Golub-Kahan bidiagonalization

Consider Golub-Kahan bidiagonalization (GK) of $A$ in the form

$$
\begin{aligned}
& w_{0}=0, \quad s_{1}=b / \beta_{1}, \quad \text { where } \quad \beta_{1}=\|b\|_{2}, \\
& \text { for } \quad j=1,2,3, \ldots \\
& \quad \alpha_{j} w_{j}=A^{T} s_{j}-\beta_{j} w_{j-1}, \quad\left\|w_{j}\right\|=1 \\
& \quad \beta_{j+1} s_{j+1}=A w_{j}-\alpha_{j} s_{j}, \quad\left\|s_{j+1}\right\|=1 \\
& \text { end. }
\end{aligned}
$$

Denote $S_{k}=\left[u_{1}, \ldots, s_{k}\right], W_{k}=\left[w_{1}, \ldots, w_{k}\right]$ resulting matrices with orthonormal columns and

$$
L_{k}=\left[\begin{array}{cccc}
\alpha_{1} & & & \\
\beta_{2} & \alpha_{2} & & \\
& \ddots & \ddots & \\
& & \beta_{k} & \alpha_{k}
\end{array}\right], \quad L_{k+}=\left[\begin{array}{c}
L_{k} \\
e_{k}^{T} \beta_{k+1}
\end{array}\right]
$$

Regularization methods based on GK compute the solution in two steps. First the problem is projected on the Krylov subspace using $k$ steps of bidiagonalization, i.e.

$$
A W_{k}=S_{k+1} L_{k+}
$$

Then an inner regularization is applied to the projected problem

$$
A x \approx b \longrightarrow L_{k+} y \approx \beta_{1} e_{1}
$$

When to stop the bidiagonalization?

## Core reduction:

From the core theory [Paige, Strakoš - 06] it follows that there exists a fundamental decomposition

$$
P^{T}[b \mid A Q]=\left[\begin{array}{c||c|c}
b_{1} & A_{11} & 0 \\
\hline 0 & 0 & A_{22}
\end{array}\right], \quad P^{T}=P^{-1}, \quad Q^{T}=Q^{-1}
$$

yielding a subproblem

$$
A_{11} x_{1} \approx b_{1}
$$

which contains all necessary and sufficient information to solve the original problem,

$$
x=Q\left[\begin{array}{c}
x_{1} \\
0
\end{array}\right]
$$

Computation of the core problem:

If the GK bidiagonalization of $A$ with $u_{1} \equiv b / \beta_{1}, \beta_{1} \equiv\|b\|$ stops with

- $\beta_{p+1}=0$ or $p=n$, then $S_{p}^{T}\left[b, A W_{p}\right]=\left[\beta_{1} e_{1}, L_{p}\right] \equiv\left[b_{1}, A_{11}\right]$ and

$$
L_{p} y=\beta_{1} e_{1}
$$

is the core problem;

- $\alpha_{p+1}=0$ or $p=m$, then $S_{p+1}^{T}\left[b, A W_{p}\right]=\left[\beta_{1} e_{1}, L_{p+}\right] \equiv\left[b_{1}, A_{11}\right]$ and

$$
L_{p+} y \approx \beta_{1} e_{1}
$$

is the core problem.

## In exact arithmetic

GK stops with $\beta_{p+1}=0$ or $\alpha_{p+1}=0$. Then the bidiagonal problem

$$
L_{p} y=\beta_{1} e_{1} \quad \text { or } \quad L_{p+} y \approx \beta_{1} e_{1}
$$

contains all necessary and sufficient information to solve the original problem.

## In floating-point arithmetic

we have to stop GK by using some stopping criteria. We can view the stopping as a perturbation of the bidiagonal matrix

$$
\beta_{k+1} \longrightarrow \widetilde{\beta}_{k+1}=0 \quad \text { or } \quad \alpha_{k+1} \longrightarrow \tilde{\alpha}_{k+1}=0
$$

which yields the modified matrix $\tilde{L}_{k}$ or $\tilde{L}_{k+}$, respectively.
How to define the stopping criteria for GK if the ill-posed problem with a noisy right-hand side is considered?

## 3. How to identify the noise

GK starts with the normalized noisy right-hand side $s_{1}=b /\|b\|$, thus vectors $s_{j}$ has to contain some information about the noise.

Our idea is: An information about the noise level can be obtained by Fourier analysis of the vectors $s_{j}$ generated by GK.

We used two different Fourier basis:

- basis of the left singular vectors $\tilde{u}_{j}$ of $A$ (basis useful for the theoretical analysis but not in practical computations);
- trigonometric basis (well applicable in practical computations, e.g., the fast Fourier transform algorithm - FFT).


## An example:

Consider the problem SHAW(400) from [Hansen, RTools] with a noisy right-hand side (the noise was artificially added)

$$
46.6225=\left\|b^{\text {exact }}\right\| \gg\left\|b^{n o i s e}\right\|=10^{-12} .
$$

We study the noise-contaminated vectors $s_{j}$ in the noise-free basis $\tilde{U}=\left[\tilde{u}_{1}, \ldots, \tilde{u}_{n}\right]$ and in the frequency domain,

$$
\left(\tilde{U}^{T} s_{j}\right) \quad \text { and } \quad F\left[s_{j}\right], \quad j=1,2, \ldots,
$$

where $F$ denotes the FFT operator.

The vector $s_{1}$ is dominated by low frequencies, thus it has dominant projection in the direction of the left singular vector $\tilde{u}_{1}$ and possibly several next vectors. Analogously $s_{2}, s_{3}, \ldots$.


For some index $j=k$ the low frequencies information is projected out from $s_{k}$ by orthogonalization against the previous vectors $s_{j}$, $j=1,2,3, \ldots, k-1$, and the noise is revealed.

$s_{6}$






$s_{19}$


Vector $s_{18}$ is fully dominated by noise -
the noise level is revealed.
Now we get explicit information when the noise begins to cover useful information in the data. The solution of the original problem $A x \approx b$ computed through the bidiagonal problem

$$
L_{j+} y \approx \beta_{1} e_{1}
$$

for $j>k=18$ can be significantly polluted by the noise.
(In the 19th step, the noise is partially projected out because vectors $s_{j}$ has to be mutually orthonormal.)

## 4. Summary and future work

Information about the noise can be obtained directly from the GolubKahan bidiagonalization.

Opened questions:

- How to implement this idea as a stopping criterion in hybrid methods?
- Relationship to common stopping criteria in hybrid methods?


## Thank you for your attention!

## References

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